

The lifting scheme of 4-channel orthogonal wavelet transforms*

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Abstract The 4-channel smooth wavelets with linear phase and orthogonality are designed from the 2-channel orthogonal wavelets with high transfer vanishing moments. Reversely, for simple lifting scheme of such 4-channel orthogonal wavelet transforms, a new 2-channel orthogonal wavelet associated with this 4-channel wavelet is constructed. The new 2-channel wavelet has at least the same number of vanishing moments as the associated 4-channel one. Finally, by combining the two such 2-channel wavelet systems, the lifting scheme of 4-channel orthogonal wavelet transform, which has simple structure and is easy to apply, is presented.

Keywords: lifting scheme, 4-channel orthogonal wavelet, 2-channel orthogonal wavelet, vanishing moment, transfer vanishing moment, linear phase.

The lifting scheme has been used in both design^[1,2] and factorization^[3] of filter banks, which is reversible and adopted in-place computation^[3]. The lifting scheme can realize fast implementations of the discrete wavelet transform, and design transform that map integers to integers^[4,5]. Chen et al.^[6] extended Sweldens' conventional 2-channel lifting scheme and presented the M -channel lifting scheme factorization of perfect reconstruction filter banks. However, this M -channel lifting scheme is complicated and difficult to apply. The aim of this paper is to simplify the lifting scheme of a special kind of 4-channel wavelet transforms.

In order to construct Heisenberg FIR (finite impulse response) filter banks, Jawerth and Peng^[7] proposed a new method of designing 4-channel symmetric low-pass filters. Peng et al.^[8] extended that method to design 2^n -channel wavelet systems with beautiful structure from the 2-channel ones. However, the vanishing moments of the designed 4-channel wavelets are incompatible with that of the associated 2-channel ones (see Fig. 1 in Ref. [8]), which is an obstacle to further study of such 4-channel wavelet transforms.

In this paper, the incompatibility is proved theoretically. Then, the concept of transfer vanishing moments of 2-channel wavelets is introduced. Hence, the 2-channel orthogonal wavelet with P transfer vanishing moments will directly generate a symmetric and orthogonal 4-channel wavelet with P vanishing

moments. To simplify the lifting scheme of such 4-channel wavelet transform, a new 2-channel orthogonal wavelet is constructed, which has at least the same vanishing moments as the 4-channel wavelet. Finally, the 4-channel orthogonal wavelet transform is divided into two independent 2-channel transforms. By combining such 2-channel wavelet transforms, the lifting scheme of 4-channel orthogonal wavelet transform is presented.

1 Relationship between 2-channel wavelets and 4-channel wavelets

A 2-channel low-pass FIR filter is denoted by $\{f_k\}$. Let $z = e^{i\omega}$, then its z -transform is given by $F(z) = \sum_{k \in \mathbb{Z}} f_k z^k$. Its polyphase decomposition is $F(z) = F_0(z^2) + zF_1(z^2)$, with $F_i(z) = \sum_{k \in \mathbb{Z}} f_{2k+i} z^k$ ($i = 0, 1$) being its two polyphase components.

Similarly, suppose the 4-channel low-pass filter is denoted by $\{h_k\}$, and its z -transform is $H(z) = \sum_{k \in \mathbb{Z}} h_k z^k$. The polyphase decomposition of h is $H(z) = H_0(z^4) + zH_1(z^4) + z^2H_2(z^4) + z^3H_3(z^4)$, with $H_i(z) = \sum_{k \in \mathbb{Z}} h_{4k+i} z^k$, ($i = 0, 1, 2, 3$) being its polyphase components. Denote by $\{g_k^i\}$ the high-pass filters for ($i = 1, 2, 3$), and $G^i(z)$ their z -transforms.

We firstly recall the algorithm in Ref. [8].

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Lemma 1 (Theorem 1 [8]). Suppose $f = (f_0, f_1, f_2, \dots, f_{2L-1})$, and its polyphase decomposition is $F(z) = F_0(z^2) + zF_1(z^2)$. Define $H(z) = F_0(z^4) + zF_1(z^4) + z^{4L-1}(F_0(z^{-4}) + z^{-1}F_1(z^{-4}))$. Then f is a 2-channel low-pass filter with length $2L$, if and only if $H(z)$ determines a 4-channel symmetric low-pass filter with length $4L$.

From Lemma 1, we know $F_i(z) = \sum_{k=0}^L f_{2k+iz^k}$ for $i = 0, 1$, and the polyphase components of the corresponding 4-channel low-pass filter are $H_i(z) = F_i(z)$, $H_{3-i}(z) = z^{L-1}F_i(z^{-1})$ with $i = 0, 1$. For the 4-channel system here is orthogonal, its polyphase matrix $\mathbf{H}_p(z)$ ^[8] should be paraunitary. Next, the simple version of $\mathbf{H}_p(z)$ ^[8] is presented by proper phase-shift.

Theorem 1. Suppose that $f = (f_0, f_1, f_2, \dots, f_{2L-1})$ is a 2-channel filter with length $2L$. Its two polyphase components are $F_0(z)$ and $F_1(z)$. When L is odd, denote $H_0(z) = z^{-\frac{L-1}{2}}F_0(z)$ and $H_1(z) = z^{-\frac{L-1}{2}}F_1(z)$. And when L is even, denote $H_0(z) = z^{-\frac{L-2}{2}}F_0(z)$ and $H_1(z) = z^{-\frac{L-2}{2}}F_1(z)$. Then $H(z) = H_0(z^4) + zH_1(z^4) + z^2H_1(z^{-4}) + z^3H_0(z^{-4})$ determines a 4-channel symmetric orthogonal low-pass filter with length $4L$, if and only if f is a 2-channel orthogonal low-pass filter.

By Theorem 1, we know that when L is either odd or even, the polyphase matrix \mathbf{H}_p can be written as

$$\mathbf{H}_p = \begin{bmatrix} H_0(z) & H_1(z) & H_1(z^{-1}) & H_0(z^{-1}) \\ -H_0(z) & H_1(z) & -H_1(z^{-1}) & H_0(z^{-1}) \\ -H_1(z) & -H_0(z) & H_0(z^{-1}) & H_1(z^{-1}) \\ H_1(z) & -H_0(z) & -H_0(z^{-1}) & H_1(z^{-1}) \end{bmatrix}, \tag{1}$$

where four entries in each row exactly correspond to the polyphase components of $H(z)$ and $G^i(z)$, $i = 1, 2, 3$. $\mathbf{H}_p(z)$ can be proved to be paraunitary and it can determine a 4-channel orthogonal system with beautiful structure.

Theorem 2. Suppose that the 4-channel polyphase decomposition of h is $H(z) = H_0(z^4) + zH_1(z^4) + z^2H_1(z^{-4}) + z^3H_0(z^{-4})$, while that of f is denoted by $F(z) = H_0(z^2) + zH_1(z^2)$. If f satisfies $P > 1$ vanishing moments, then the number

of vanishing moment of h is less than 2.

Proof. If f satisfies $P > 1$ vanishing moments, at least $P = 2$, then by Ref. [10], $\left. \frac{dF}{d\omega} \right|_{\omega=\pi} = 0$ holds.

Expanding it, we have

$$\frac{dF}{d\omega} = 2iz^2 \left\{ \frac{dH_0}{dz}(z^2) + z \frac{dH_1}{dz}(z^2) \right\} + izH_1(z^2).$$

Let $H_i^1(z) = 4z \frac{dH_i}{dz}(z)$ for $i = 0, 1$ and $\omega = \pi$.

Then we have $H_0^1(1) - H_1^1(1) = 2$. Suppose that h has $P = 2$ vanishing moments, then $\left. \frac{dH}{d\omega} \right|_{\omega=\{\frac{\pi}{2}, \pi, \frac{3\pi}{2}\}} = 0$ holds.

Expanding it, we have

$$\begin{aligned} \frac{dH}{d\omega} = i \{ & [H_0^1(z^4) + zH_1^1(z^4) \\ & - z^2H_1^1(z^{-4}) - z^3H_0^1(z^{-4})] \\ & + [zH_1(z^4) + 2z^2H_1(z^{-4}) \\ & + 3z^3H_0(z^{-4})] \}. \end{aligned}$$

Let $\omega \in \left\{ \frac{\pi}{2}, \pi, \frac{3\pi}{2} \right\}$, then $H_0^1(1) = \frac{3}{2}$ and $H_1^1(1) = \frac{1}{2}$ hold. It is in contradiction. Therefore, if f has arbitrary $P > 1$ vanishing moments, then h satisfies at most $P = 1$ vanishing moment. Q. E. D.

Theorem 2 shows that the vanishing moment of h is incompatible with that of f . For designing smooth 4-channel wavelets, new conditions should be imposed on $F(z)$. Hence, the concept of transfer vanishing moment is introduced. Using the definition in the proof of Theorem 2, let $H_i^1(z) = 4z \frac{dH_i}{dz}(z)$ with $i = 0, 1$. If $F(z) = H_0(z^2) + zH_1(z^2)$ satisfies the equations corresponding to $p = 0, 1, \dots, P - 1$ listed in Table 1, then $F(z)$ is called satisfying P transfer vanishing moments.

Table 1. Transfer vanishing moments

p	H_0^p	H_1^p
0	$H_0^0(1) = 1$	$H_1^0(1) = 1$
1	$H_0^1(1) = \frac{3}{2}$	$H_1^1(1) = \frac{1}{2}$
2	$H_0^2(1) = H_1^2(1) + 2$	—
3	$H_0^3(1) = \frac{9}{2}H_0^1(1) - \frac{27}{4}$	$H_1^3(1) = \frac{3}{2}H_1^1(1) - \frac{1}{4}$
4	$H_0^4(1) = H_1^4(1) + 12H_1^2(1) + 2$	—
...

Theorem 3. Suppose that $H_i(z)$ ($i = 0, 1$) is defined in Theorem 1. $H(z) = H_0(z^4) + zH_1(z^4) + z^2H_1(z^{-4}) + z^3H_0(z^{-4})$ determines 4-channel

wavelets with P vanishing moments, if and only if $F(z) = H_0(z^2) + zH_1(z^2)$ determines 2-channel wavelets with P transfer vanishing moments.

To construct 4-channel orthogonal wavelet with P vanishing moments, it is only necessary to design a 2-channel orthogonal wavelet that has P transfer vanishing moments. After the smooth 4-channel wavelets are designed, we then should simplify their transforms. Since the transfer vanishing moment is much different from the traditional vanishing moment, we introduce a new 2-channel low-pass filter \tilde{f} associated with h . It can be proved that this filter has the same vanishing moments as h . The polyphase decomposition of \tilde{f} is

$$\tilde{F}(z) \triangleq H_0(z^2) + zH_1(z^{-2}). \quad (2)$$

Clearly, (i) $\tilde{F}(z)$ is the combination of the first component with the third one of $H(z)$; (ii) $\tilde{F}(z)$ also determines the 2-channel orthogonal filter \tilde{f} ; (iii) \tilde{f}_k , each entry of \tilde{f} , corresponds to h_{2k} .

In the following part, we will discuss the compatibility of the 2-channel and 4-channel systems in the vanishing moments.

As for the 2-channel low-pass filter $F(z) = F_0(z^2) + zF_1(z^2)$, we have

Lemma 2. Let $F_i^0(z) = F_i(z)$ ($i = 0, 1$), and $F_i^n(z) = 2z \frac{dF_i^{n-1}}{dz}(z)$ ($i = 0, 1, n = 1, 2, \dots$).

Then the following equation holds:

$$\begin{aligned} \frac{d^n F}{d\omega^n} = & i^n \left\{ [F_0^n(z^4) + zF_1^n(z^4)] \right. \\ & + \binom{n}{1} [zF_1^{n-1}(z^4)] + \dots \\ & + \binom{n}{m} [zF_1^{n-m}(z^4)] + \dots \\ & \left. + \binom{n}{n} [zF_1^0(z^4)] \right\}. \quad (3) \end{aligned}$$

Similarly, as for the general 4-channel low-pass filter $H(z) = H_0(z^4) + zH_1(z^4) + z^2H_2(z^4) + z^3H_3(z^4)$, we have

Lemma 3. Let $H_i^0(z) = H_i(z)$ ($i = 0, 1, 2, 3$), and $H_i^n(z) = 4z \frac{dH_i^{n-1}}{dz}(z)$ ($i = 0, 1, 2, 3, n = 1, 2, \dots$). Then

$$\begin{aligned} \frac{d^n H}{d\omega^n} = & i^n \left\{ [H_0^n(z^4) + zH_1^n(z^4)] \right. \\ & + z^2H_2^n(z^4) + z^3H_3^n(z^4)] \\ & + \binom{n}{1} [zH_1^{n-1}(z^4) + 2z^2H_2^{n-1}(z^4) \\ & + 3z^3H_3^{n-1}(z^4)] + \dots \\ & + \binom{n}{m} [zH_1^{n-m}(z^4) + 2^m z^2H_2^{n-m}(z^4) \\ & + 3^m z^3H_3^{n-m}(z^4)] + \dots \\ & \left. + \binom{n}{n} [zH_1^0(z^4) + 2^n z^2H_2^0(z^4) \right. \\ & \left. + 3^n z^3H_3^0(z^4)] \right\}. \quad (4) \end{aligned}$$

For the special relation between $F_i(z)$ and $H_i(z)$ in Eq. (2), substituting $\tilde{F}(z)$ into Eq. (3) and Eq. (4), we have

$$\frac{d^n \tilde{F}}{d\omega^n} \Big|_{\omega=\pi} = \frac{1}{2^n} \left(\frac{d^n H}{d\omega^n} \Big|_{\omega=\frac{\pi}{2}} + \frac{d^n H}{d\omega^n} \Big|_{\omega=\frac{3\pi}{2}} \right).$$

Clearly, if h satisfies $\forall P$ vanishing moments, \tilde{f} corresponding to Eq. (2) satisfies at least P vanishing moments.

Fig.1 shows the comparison of scaling functions. Fig.1(a) is the scaling function Φ corresponding to the 4-channel low-pass filter h with length 20 designed in Ref. [8], while Fig.1(b) and Fig.1(c) show the scaling functions ϕ and $\tilde{\phi}$ corresponding to f and \tilde{f} , respectively. Clearly, $\tilde{\phi}$ is smooth, but ϕ is not.

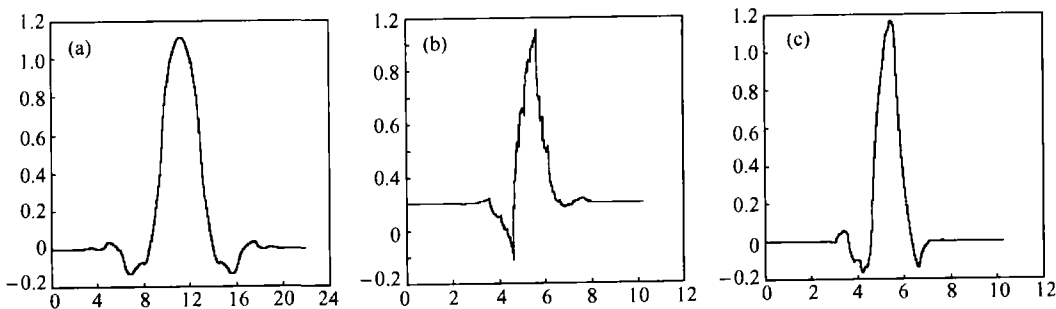


Fig. 1. Comparison of scaling functions.

2 4-channel lifting scheme with simple structure

Because $H_p(z)$ is paraunitary, the 4-channel wavelet system determined by Eq. (1) is orthogonal. As for an input signal x , $X(z) = X_0(z^4) + zX_2(z^4) + z^2X_1(z^4) + z^3X_3(z^4)$ is its polyphase decomposition, where $X_i(z)$ with $i = 0, 1, 2, 3$ are its polyphase components. Letting $\mathbf{X}(z) = [X_0(z), X_2(z), X_1(z), X_3(z)]^T$, the 4-channel wavelet transform can be expressed by $H_p(z)\mathbf{X}(z)$.

Transposing the second column with the third one of H_p , we have

$$\tilde{H}_p = \begin{bmatrix} H_0(z) & H_1(z^{-1}) & H_1(z) & H_0(z^{-1}) \\ -H_0(z) & -H_1(z^{-1}) & H_1(z) & H_0(z^{-1}) \\ -H_1(z) & H_0(z^{-1}) & -H_0(z) & H_1(z^{-1}) \\ H_1(z) & -H_0(z^{-1}) & -H_0(z) & H_1(z^{-1}) \end{bmatrix} \quad (5)$$

For clear discussion, \tilde{H}_p is written as the form of blocked matrix. Then, the 4-channel wavelet transform becomes

$$H_p(z)\mathbf{X}(z) = \tilde{H}_p(z) \cdot [X_0(z) X_2(z) X_1(z) X_3(z)]^T \quad (6)$$

In the expression of \tilde{H}_p , the two entries in each row of each block correspond to the two polyphase

components of one 2-channel filter, either low-pass or high-pass. If the 2-channel polyphase decomposition of the input signal x is written as $X(z) = X^0(z^2) + zX^1(z^2)$, then $X^0(z) = X_0(z^2) + zX_2(z^2)$ and $X^1(z) = X_1(z^2) + zX_3(z^2)$ hold. Therefore, Eq. (6) can be rewritten as

$$H_p(z)\mathbf{X}(z) = \begin{bmatrix} (\downarrow 2)\tilde{F}(z) & (\downarrow 2)\bar{F}(z) \\ (\downarrow 2)(-\tilde{F}(z)) & (\downarrow 2)\bar{F}(z) \\ (\downarrow 2)\tilde{G}(z) & (\downarrow 2)\bar{G}(z) \\ (\downarrow 2)(-\tilde{G}(z)) & (\downarrow 2)\bar{G}(z) \end{bmatrix} \cdot \begin{bmatrix} X^0(z) \\ X^1(z) \end{bmatrix}, \quad (7)$$

where $\tilde{F}(z)$ is exactly defined by Eq. (2), and $\bar{F} = H_1(z^2) + zH_0(z^{-2})$, while $\tilde{G}(z)$ and $\bar{G}(z)$ are the high-pass filters corresponding to $\tilde{F}(z)$ and $\bar{F}(z)$, respectively. Also, $\bar{F}(z) = z^{-1}\tilde{F}(z^{-1})$ holds, which shows $\tilde{\phi}$ and $\bar{\phi}$ have the same regularity.

From Eq. (7), the 4-channel orthogonal wavelet transform can be simply divided into two independent 2-channel wavelet transforms. Fig. 2 shows two equal structures of 4-channel wavelet transforms. Fig. 2(a) gives the general 4-channel wavelet transform, with $H(z)$ and $G^i(z)$ being the low-pass filter and three high-pass filters for $i = 1, 2, 3$. Fig. 2(b) presents the combination of two 2-channel wavelet transforms in Eq. (7), each of which is shown in one dash frame respectively.

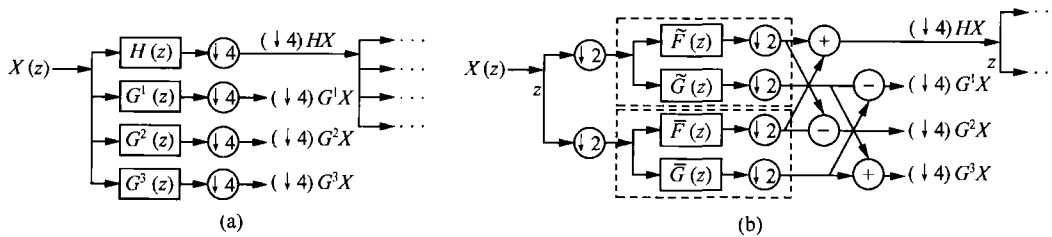


Fig. 2. Two equal structures of 4-channel wavelet transforms.

The upper dash frame in Fig. 2(b) shows the 2-channel orthogonal wavelet transform based on $\tilde{F}(z)$ and $\tilde{G}(z)$, and the lower one shows that based on $\bar{F}(z)$ and $\bar{G}(z)$. Therefore, lifting scheme of 4-channel system can be simply divided into two 2-channel lifting schemes. The lifting scheme of Eq. (7) is shown in Fig. 3, where $\bar{s}_i(z)$, $\bar{t}_i(z)$, $\tilde{s}_i(z)$ and $\tilde{t}_i(z)$ represent the lifting steps of the two 2-channel systems, respectively.

Using the factorization algorithm in Ref. [3], we give the factorization of $\tilde{F}(z)$, whose scaling function is shown in Fig. 1(c):

$$\begin{bmatrix} H_0(z) \\ H_1(z^{-1}) \end{bmatrix} = \begin{bmatrix} 1 & \alpha_{00} \\ 0 & 1 \end{bmatrix} \cdot \begin{bmatrix} 1 & 0 \\ \alpha_{10} + \alpha_{11}z^{-1} & 1 \end{bmatrix} \cdot \begin{bmatrix} 1 & \alpha_{20} + \alpha_{21}z \\ 0 & 1 \end{bmatrix} \cdot \begin{bmatrix} 1 & 0 \\ \alpha_{30} + \alpha_{31}z^{-1} & 1 \end{bmatrix} \cdot \begin{bmatrix} 1 & \alpha_{40} + \alpha_{41}z \\ 0 & 1 \end{bmatrix} \cdot \begin{bmatrix} 0 \\ K \end{bmatrix}. \quad (8)$$

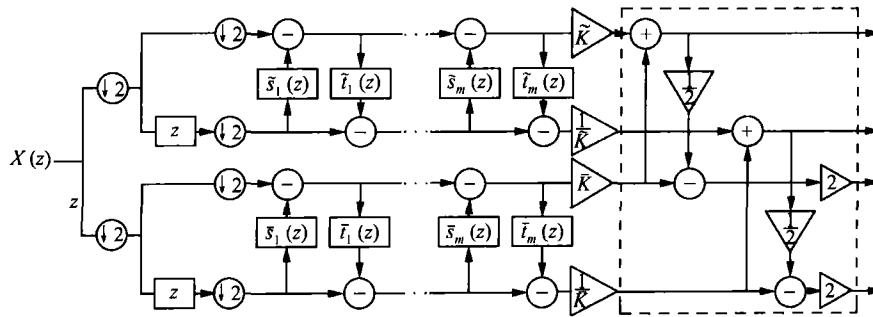


Fig. 3. 4-channel lifting scheme with simple structure.

The parameters are $(\alpha_{00}, \alpha_{10}, \alpha_{11}, \alpha_{20}, \alpha_{21}, \alpha_{30}, \alpha_{31}, \alpha_{40}, \alpha_{41}) = (-5.3836482, -0.009215727, 0.1795527, -28.043056, 7.6974209, 0.0022974888, 0.017948667, -18.39089, -7.1095098, -0.12772164)$.

Ref. [4] gives two solutions to convert the scaling operation (K or $\frac{1}{K}$ in Fig. 3) into maps from integers into integers. And the operation in the dash frame in Fig. 3 can also map integers into integers. Therefore, 4-channel lifting scheme can also be converted into the map from integers into integers.

3 Conclusion

In this paper, smooth symmetric 4-channel orthogonal wavelets have been designed by finding the 2-channel orthogonal wavelets that satisfy high transfer vanishing moments. The concept of transfer vanishing moment is introduced, because the vanishing moment of the 2-channel wavelets is incompatible with that of the 4-channel ones. For simple lifting scheme of such 4-channel orthogonal wavelet transforms, a new 2-channel orthogonal wavelet is constructed, which satisfies at least the same vanishing moment as the 4-channel wavelet. By combining such 2-channel wavelet transforms, a simplified lifting

scheme of 4-channel orthogonal transform is given.

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